LOW DROPOUT VOLTAGE REGULATOR

## FEATURES

- **High Voltage Precision at ± 2.0% or ± 60 mV**
- **Very Low Quiescent Current**
- **Very Low Dropout Voltage**
- **Reverse Voltage Protection**
- **Miniature Package (SOT23-3)**
- **Short Circuit Protection**
- **High Ripple Rejection**
- **Can use Multilayer Ceramic Capacitors**

## APPLICATIONS

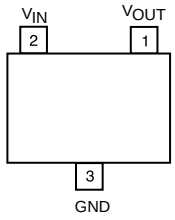
- **Battery Powered Systems**
- **Cellular Telephones**
- **Pagers**
- **Personal Communications Equipment**
- **Portable Instrumentation**
- **Portable Consumer Equipment**
- **Radio Control Systems**
- **Toys**
- **Low Voltage Systems**

## DESCRIPTION

The TK715xx is a low dropout linear regulator housed in a small SOT23-3 package, rated at 400 mW. An internal PNP transistor is used to achieve a low dropout voltage of 105 mV (typ.) at 50 mA load current. This device offers high precision output voltage of ± 2.0 % or ± 60 mV. The TK715xx has a very low quiescent current of 25 µA (typ.) at no load. The low quiescent current and dropout voltage make this part ideal for battery powered applications. The internal reverse bias protection eliminates the requirement for a reverse voltage protection diode, saving cost and board space. The high 60 dB ripple rejection and low noise provide enhanced performance for critical applications.

**TK715xxAS**




## ORDERING INFORMATION

TK715□□ASCL

- Voltage Code
- Tape/ Reel Code
- Temp. Code
- Package Code

VOLTAGE CODE

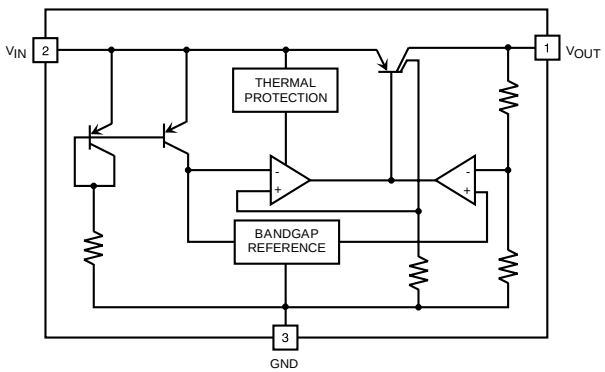
| | |
|---|---|
| 15 = 1.5 V | 35 = 3.5 V |
| 16 = 1.6 V | 36 = 3.6 V |
| 17 = 1.7 V | 37 = 3.7 V |
| 18 = 1.8 V | 38 = 3.8 V |
| 19 = 1.9 V | 39 = 3.9 V |
| 20 = 2.0 V | 40 = 4.0 V |
| 21 = 2.1 V | 41 = 4.1 V |
| 22 = 2.2 V | 42 = 4.2 V |
| 23 = 2.3 V | 43 = 4.3 V |
| 24 = 2.4 V | 44 = 4.4 V |
| 25 = 2.5 V | 45 = 4.5 V |
| 26 = 2.6 V | 46 = 4.6 V |
| 27 = 2.7 V | 47 = 4.7 V |
| 28 = 2.8 V | 48 = 4.8 V |
| 29 = 2.9 V | 49 = 4.9 V |
| 30 = 3.0 V | 50 = 5.0 V |
| 31 = 3.1 V | 60 = 6.0 V |
| 32 = 3.2 V | 70 = 7.0 V |
| 33 = 3.3 V | 80 = 8.0 V |
| 34 = 3.4 V | 90 = 9.0 V |

TEMPERATURE CODE
C -30 to +80 °C

PACKAGE CODE
S : SOT-23-3

TAPE/REEL CODE
L: Tape Left

**BLOCK DIAGRAM**



## ABSOLUTE MAXIMUM RATINGS ($V_{OUT} \leq 5.0$ V)

Supply Voltage ..... -0.4 to 19 V
Power Dissipation (Note 1) ..... 400 mW
Reverse Bias ..... 8 V
Short Circuit Current ..... 170 mA
Storage Temperature (Ambient) ..... -55 to +150 °C
Operating Temperature (Ambient) ..... -30 to +80 °C
Max. Operating Temperature (Junction) ..... 125 °C
Operating Voltage Range ..... 1.8 to 18.0 V
Junction Temperature ..... 150 °C
Lead Soldering Temperature (10 s) ..... 235 °C

## TK715xx ELECTRICAL CHARACTERISTICS ($V_{OUT} \leq 5.0$ V)

Test conditions: $V_{IN} = V_{OUT(TYP)} + 1V$, $T_A = 25$ °C, unless otherwise specified.

| SYMBOL | PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNITS |
|---|---|---|---|---|---|---|
| $I_Q$ | Quiescent Current | $I_{OUT} = 0$ mA | | 25 | 45 | µA |
| $I_{GND}$ | Ground Pin Current | $I_{OUT} = 15$ mA | | 300 | 500 | µA |
| $V_{OUT}$ | Output Voltage | $I_{OUT} = 5$ mA | See Table 1 | | | V |
| Line Reg | Line Regulation | $V_{IN} = V_{OUT(TYP)} + 1$ V to $V_{OUT(TYP)} + 6$ V | | 3 | 12 | mV |
| Load Reg | Load Regulation | $I_{OUT} = 5$ to 100 mA, (Note 2) | | 18 | 36 | mV |
| $V_{DROP}$ | Dropout Voltage (Note 5) | $I_{OUT} = 50$ mA | | 0.105 | 0.18 | V |
| | | $I_{OUT} = 100$ mA, 2.4 V $\leq V_{OUT} \leq$ 5.0 V | | 0.16 | 0.28 | V |
| | | $I_{OUT} = 100$ mA, 2.1 V $\leq V_{OUT} \leq$ 2.4 V | | 0.16 | 0.30 | V |
| $I_{OUT(MAX)}$ | Continuous Output Current | | 115 | 155 | | mA |
| | | 1.8 V $\leq V_{in} \leq$ 2.1 V (Note 3) | 70 | 90 | | |
| RR | Ripple Rejection | (Note 4) | | 60 | | dB |
| $\Delta V_{OUT} / \Delta T$ | Temperature Coefficient | $I_{OUT} = 5$ mA | | 30 | | ppm/°C |

Note 1: Power dissipation is 400 mW when mounted as recommended. Derate at 3.2 mW/°C for operation above 25 °C.
Note 2: Refer to "Definition of Terms."
Note 3: Please refer to the Applications Section for more information.
Note 4: Ripple rejection is measured at $V_R = 200$ mVrms, $V_{IN} = V_{OUT(TYP)} + 2$ V, $I_{OUT} = 10$ mA, $C_L = 2.2$ µF, f = 100 Hz.
Note 5: The minimum operating voltage for $V_{IN}$ can be 1.8 V. Also, the minimum voltage required for $V_{IN}$ is $V_{IN} = V_{DROP} + V_{OUT}$. As a result, operating at $V_{OUT} \leq 2.0$ V at the minimum input operating voltage is not preferred.
Gen. Note: Parameters with min. or max. values are 100% tested at $T_A = 25$ °C.

## ABSOLUTE MAXIMUM RATINGS ($V_{OUT} \geq 5.1$ V)

Supply Voltage ..... -0.4 to 19 V
Power Dissipation (Note 1) ..... 400 mW
Reverse Bias ..... 8 V
Short Circuit Current ..... 170 mA
Storage Temperature (Ambient) ..... -55 to +150 °C
Operating Temperature (Ambient) ..... -30 to +80 °C
Max. Operating Temperature (Junction) ..... 125 °C
Operating Voltage Range ..... 1.8 to 18 V
Junction Temperature ..... 150 °C
Lead Soldering Temperature (10 s) ..... 235 °C

## TK715xx ELECTRICAL CHARACTERISTICS ($V_{OUT} \geq 5.1$ V)

Test conditions: $V_{IN} = V_{OUT(TYP)} + 1V$, $T_A = 25$ °C, unless otherwise specified.

| SYMBOL | PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNITS |
|---|---|---|---|---|---|---|
| $I_Q$ | Quiescent Current | $I_{OUT}$ = 0 mA | | 32 | 60 | µA |
| $I_{GND}$ | Ground Pin Current | $I_{OUT}$ = 15 mA | | 300 | 500 | µA |
| $V_{OUT}$ | Output Voltage | $I_{OUT}$ = 5 mA | See Table 1 | | | V |
| Line Reg | Line Regulation | $V_{IN} = V_{OUT(TYP)} + 1$ V to $V_{OUT(TYP)} + 6$ V or Max 18 V | | 3 | 12 | mV |
| Load Reg | Load Regulation | $I_{OUT}$ = 5 to 100 mA, (Note 2) | | 35 | 80 | mV |
| $V_{DROP}$ | Dropout Voltage | $I_{OUT}$ = 50 mA | | 0.105 | 0.18 | V |
| | | $I_{OUT}$ = 100 mA | | 0.160 | 0.28 | V |
| $I_{OUT(MAX)}$ | Continuous Output Current | | 115 | 155 | | mA |
| RR | Ripple Rejection | (Note 3) | | 60 | | dB |
| $\Delta V_{OUT} / \Delta T$ | Temperature Coefficient | $I_{OUT}$ = 5 mA | | 30 | | ppm/°C |

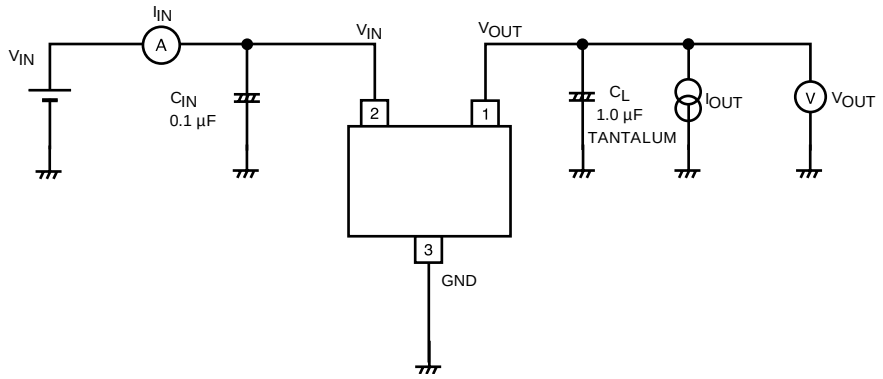
Note 1: Power dissipation is 400 mW when mounted as recommended. Derate at 3.2 mW/°C for operation above 25 °C.
Note 2: Refer to "Definition of Terms."
Note 3: Ripple rejection is measured at $V_R$ = 200 mVrms, $V_{IN} = V_{OUT(TYP)} + 2$ V, $I_{OUT}$ = 10 mA, $C_L$ = 2.2 µF, f = 100 Hz.
Gen. Note: Parameters with min. or max. values are 100% tested at $T_A$ = 25 °C.

## TK715xxAS ELECTRICAL CHARACTERISTICS TABLE 1

| Output Voltage | Voltage Code | $V_{OUT(MIN)}$ | $V_{OUT(MAX)}$ | Test Voltage |
|---|---|---|---|---|
| 1.5 V | 15 | 1.440 V | 1.560 V | 2.5 V |
| 1.6 V | 16 | 1.540 V | 1.660 V | 2.6 V |
| 1.7 V | 17 | 1.640 V | 1.760 V | 2.7 V |
| 1.8 V | 18 | 1.740 V | 1.860 V | 2.8 V |
| 1.9 V | 19 | 1.840 V | 1.960 V | 2.9 V |
| 2.0 V | 20 | 1.940 V | 2.060 V | 3.0 V |
| 2.1 V | 21 | 2.040 V | 2.160 V | 3.1 V |
| 2.2 V | 22 | 2.140 V | 2.260 V | 3.2 V |
| 2.3 V | 23 | 2.240 V | 2.360 V | 3.3 V |
| 2.4 V | 24 | 2.340 V | 2.460 V | 3.4 V |
| 2.5 V | 25 | 2.440 V | 2.560 V | 3.5 V |
| 2.6 V | 26 | 2.540 V | 2.660 V | 3.6 V |
| 2.7 V | 27 | 2.640 V | 2.760 V | 3.7 V |
| 2.8 V | 28 | 2.740 V | 2.860 V | 3.8 V |
| 2.9 V | 29 | 2.840 V | 2.960 V | 3.9 V |
| 3.0 V | 30 | 2.940 V | 3.060 V | 4.0 V |
| 3.1 V | 31 | 3.040 V | 3.160 V | 4.1 V |
| 3.2 V | 32 | 3.140 V | 3.260 V | 4.2 V |
| 3.3 V | 33 | 3.240 V | 3.360 V | 4.3 V |
| 3.4 V | 34 | 3.340 V | 3.460 V | 4.4 V |

| Output Voltage | Voltage Code | $V_{OUT(MIN)}$ | $V_{OUT(MAX)}$ | Test Voltage |
|---|---|---|---|---|
| 3.5 V | 35 | 3.440 V | 3.560 V | 4.5 V |
| 3.6 V | 36 | 3.530 V | 3.670 V | 4.6 V |
| 3.7 V | 37 | 3.630 V | 3.770 V | 4.7 V |
| 3.8 V | 38 | 3.730 V | 3.870 V | 4.8 V |
| 3.9 V | 39 | 3.830 V | 3.970 V | 4.9 V |
| 4.0 V | 40 | 3.930 V | 4.070 V | 5.0 V |
| 4.1 V | 41 | 4.030 V | 4.170 V | 5.1 V |
| 4.2 V | 42 | 4.130 V | 4.270 V | 5.2 V |
| 4.3 V | 43 | 4.230 V | 4.370 V | 5.3 V |
| 4.4 V | 44 | 4.330 V | 4.470 V | 5.4 V |
| 4.5 V | 45 | 4.430 V | 4.570 V | 5.5 V |
| 4.6 V | 46 | 4.530 V | 4.670 V | 5.6 V |
| 4.7 V | 47 | 4.630 V | 4.770 V | 5.7 V |
| 4.8 V | 48 | 4.730 V | 4.870 V | 5.8 V |
| 4.9 V | 49 | 4.830 V | 4.970 V | 5.9 V |
| 5.0 V | 50 | 4.930 V | 5.070 V | 6.0 V |
| 6.0 V | 60 | 5.880 V | 6.120 V | 7.0 V |
| 7.0 V | 70 | 6.860 V | 7.140 V | 8.0 V |
| 8.0 V | 80 | 7.840 V | 8.160 V | 9.0 V |
| 9.0 V | 90 | 8.820 V | 9.180 V | 10.0 V |

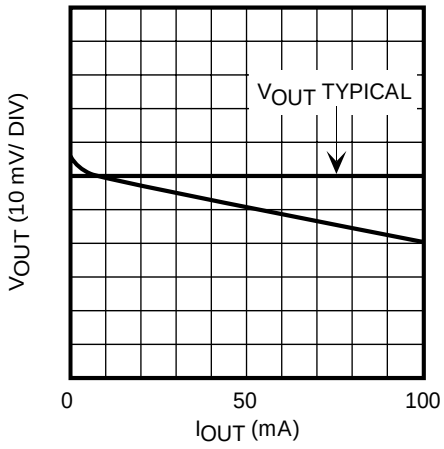
## TEST CIRCUIT



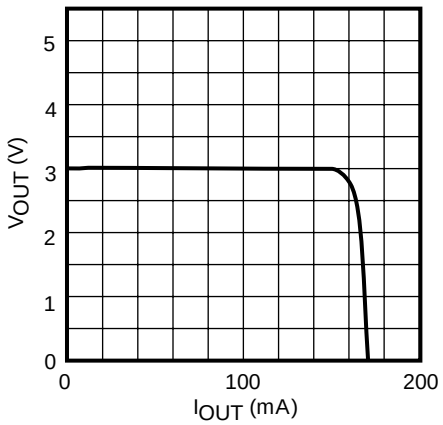

## TYPICAL PERFORMANCE CHARACTERISTICS

$T_A$ = 25 °C, unless otherwise specified.

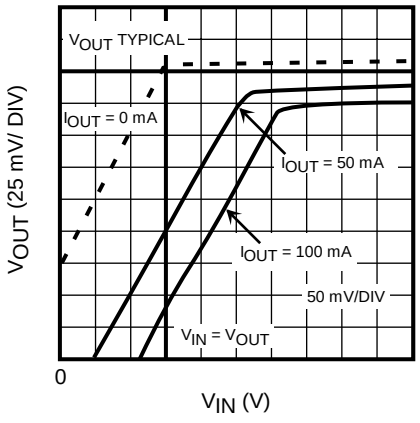
**LOAD REGULATION**



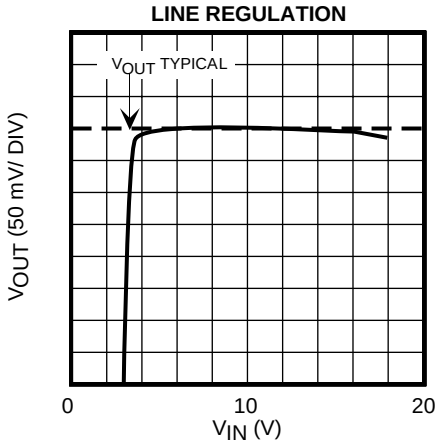

**SHORT CIRCUIT CURRENT**



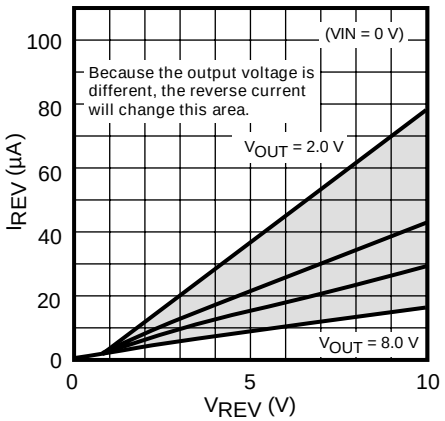

**OUTPUT VOLTAGE vs. INPUT VOLTAGE**



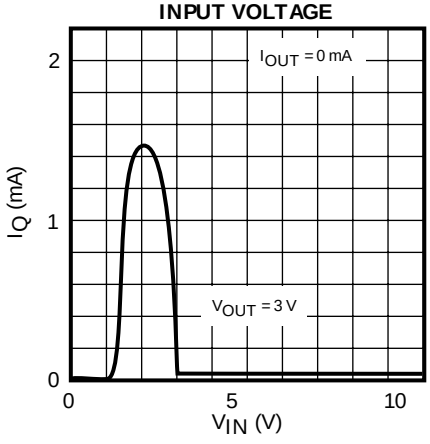

**LINE REGULATION**




**REVERSE BIAS CURRENT RANGE**



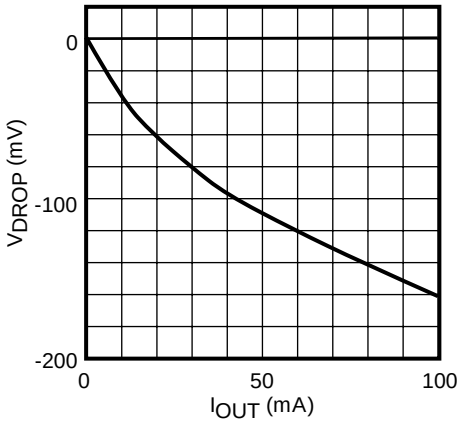

**INPUT CURRENT vs. INPUT VOLTAGE**



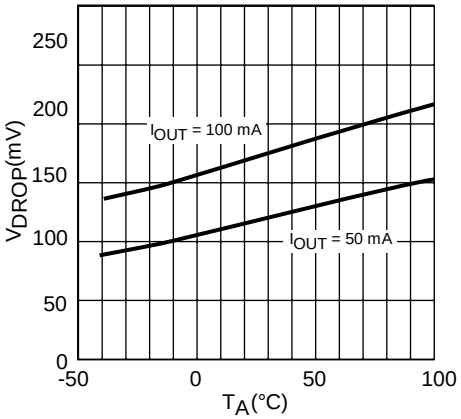
## TYPICAL PERFORMANCE CHARACTERISTICS (CONT.)

$T_A$ = 25 °C, unless otherwise specified.

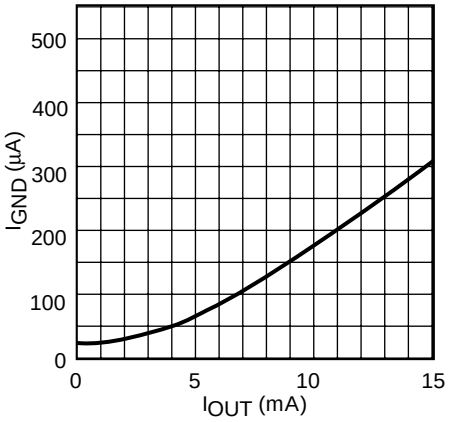
**DROPOUT VOLTAGE vs. OUTPUT CURRENT**



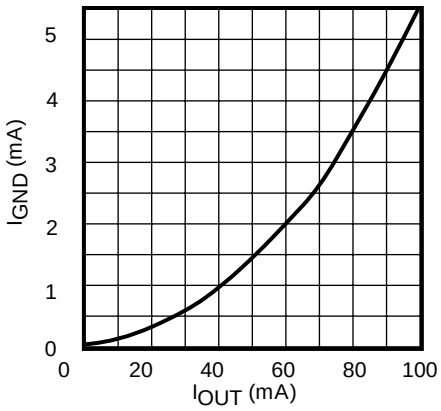

**DROPOUT VOLTAGE vs. TEMPERATURE**



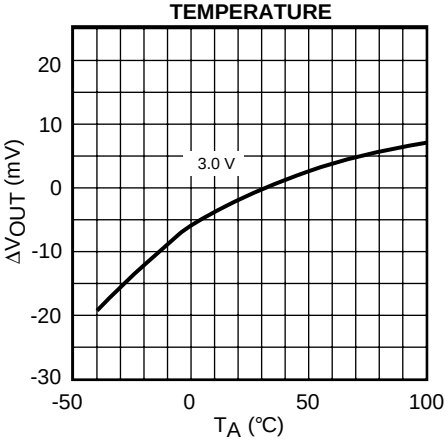

**GROUND PIN CURRENT vs. OUTPUT CURRENT**



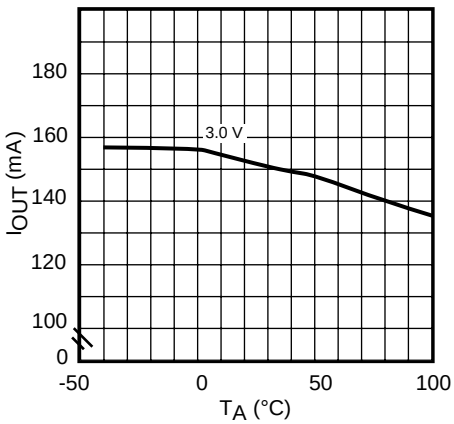

**GROUND PIN CURRENT vs. OUTPUT CURRENT**




**OUTPUT VOLTAGE vs. TEMPERATURE**

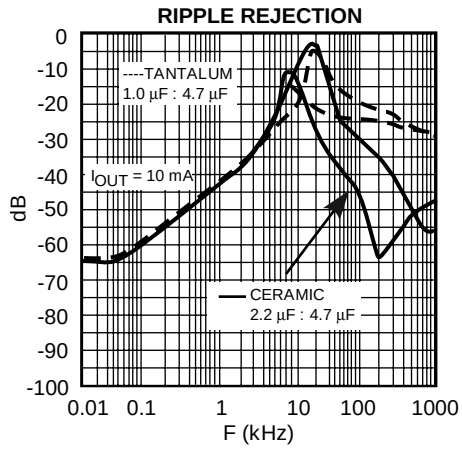


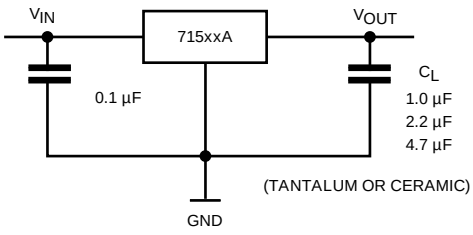

**MAX OUTPUT CURRENT**




## Ripple Rejection

**RIPPLE REJECTION**



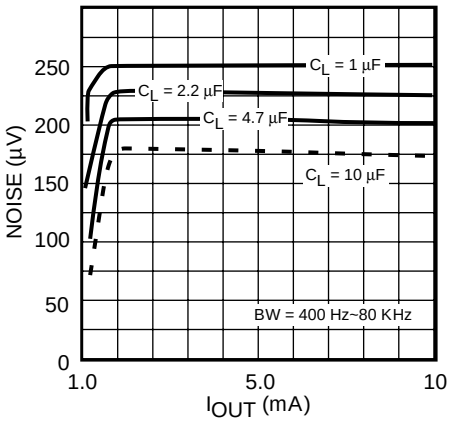

RIPPLE REJECTION CIRCUIT

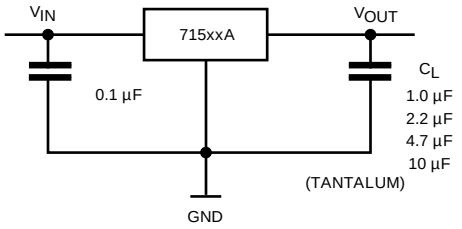
## TYPICAL PERFORMANCE CHARACTERISTICS (CONT.)

$T_A$ = 25 °C, unless otherwise specified.

### Output Noise

**NOISE LEVEL (TK71530A) vs. OUTPUT CURRENT**

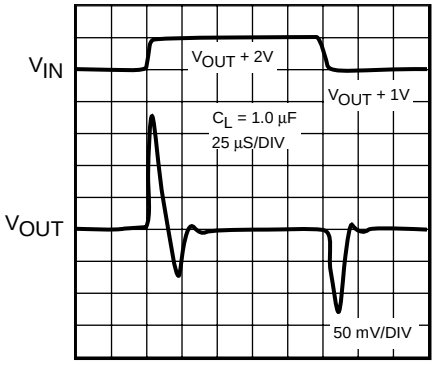



OUTPUT NOISE CIRCUIT

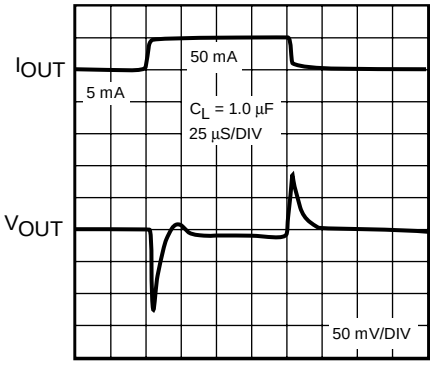
### Load & Line Response

**LINE VOLTAGE STEP RESPONSE**




$I_{OUT}$ = 10mA

**LOAD CURRENT STEP RESPONSE**




Note: To improve the load and line transient response, increase the value of the output capacitor.

## DEFINITION AND EXPLANATION OF TECHNICAL TERMS

### OUTPUT VOLTAGE ($V_{OUT}$)

The output voltage is specified with $V_{IN} = (V_{OUT(TYP)} + 1\ V)$ and $I_{OUT} = 5$ mA.

### DROPOUT VOLTAGE ($V_{DROP}$)

The dropout voltage is the difference between the input voltage and the output voltage at which point the regulator starts to fall out of regulation. Below this value, the output voltage will fall as the input voltage is reduced. It is dependent upon the load current and the junction temperature.

### CONTINUOUS OUTPUT CURRENT ($I_{OUT}$)

Normal operating output current. This is limited by package power dissipation.

### LINE REGULATION (Line Reg)

Line regulation is the ability of the regulator to maintain a constant output voltage as the input voltage changes. The line regulation is specified as the input voltage is changed from $V_{IN} = V_{OUT(TYP)} + 1\ V$ to $V_{IN} = V_{OUT(TYP)} + 6\ V$ or $V_{IN}$ = max 18 V.

### LOAD REGULATION (Load Reg)

Load regulation is the ability of the regulator to maintain a constant output voltage as the load current changes. It is a pulsed measurement to minimize temperature effects with the input voltage set to $V_{IN} = V_{OUT(TYP)} + 1\ V$. The load regulation is specified under the output current step condition 5 mA to 100 mA.

### QUIESCENT CURRENT ($I_Q$)

The quiescent current is the current which flows through the ground terminal under no load conditions ($I_{OUT} = 0$ mA).

### GROUND CURRENT ($I_{GND}$)

Ground current is the current which flows through the ground pin(s). It is defined as $I_{IN} - I_{OUT}$, excluding control current.

### RIPPLE REJECTION RATIO (RR)

Ripple rejection is the ability of the regulator to attenuate the ripple content of the input voltage at the output. It is specified with 200 mVrms, 100 Hz superimposed on the input voltage, where $V_{IN} = V_{OUT(TYP)} + 2.0\ V$. The output decoupling capacitor is set to 2.2 µF and the load current is set to 10 mA. Ripple rejection is the ratio of the ripple content of the output vs. the input and is expressed in dB.

### REVERSE VOLTAGE PROTECTION

Reverse voltage protection prevents damage due to the output voltage being higher than the input voltage. This fault condition can occur when the output capacitor remains charged and the input is reduced to zero, or when an external voltage higher than the input voltage is applied to the output side.

### REDUCTION OF OUTPUT NOISE

Although the architecture of the Toko regulators are designed to minimize semiconductor noise, further reduction can be achieved by the selection of external components. The obvious solution is to increase the size of the output capacitor. Please note that several parameters are affected by the value of the capacitors and bench testing is recommended when deviating from standard values.

### PACKAGE POWER DISSIPATION ($P_D$)

This is the power dissipation level at which the thermal sensor is activated. The IC contains an internal thermal sensor which monitors the junction temperature. When the junction temperature exceeds the monitor threshold of 150 °C, the IC is shut down. The junction temperature rises as the difference between the input power ($V_{IN} \times I_{IN}$) and the output power ($V_{OUT} \times I_{OUT}$) increases. The rate of temperature rise is greatly affected by the mounting pad configuration on the PCB, the board material, and the ambient temperature. When the IC mounting has good thermal conductivity, the junction temperature will be low even if the power dissipation is great. When mounted on the recommended mounting pad, the power dissipation of the SOT23-3 is increased to 400 mW. For operation at ambient temperatures over 25 °C, the power dissipation of the SOT23-3 device should be derated at 3.2 mW/°C. To

## DEFINITION AND EXPLANATION OF TECHNICAL TERMS (CONT.)

determine the power dissipation for shutdown when mounted, attach the device on the actual PCB and deliberately increase the output current (or raise the input voltage) until the thermal protection circuit is activated. Calculate the power dissipation of the device by subtracting the output power from the input power. These measurements should allow for the ambient temperature of the PCB. The value obtained from $P_D$ /(150 °C - $T_A$) is the derating factor. The PCB mounting pad should provide maximum thermal conductivity in order to maintain low device temperatures. As a general rule, the lower the temperature, the better the reliability of the device. The thermal resistance when mounted is expressed as follows:

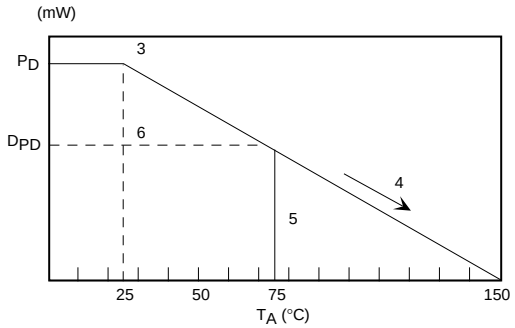
$$T_j = \theta_{jA} \times P_D + T_A$$

For Toko ICs, the internal limit for junction temperature is 150 °C. If the ambient temperature ($T_A$) is 25 °C, then:

$$150\ °C = \theta_{jA} \times P_D + 25\ °C$$
$$\theta_{jA} = 125\ °C / P_D$$

$P_D$ is the value when the thermal sensor is activated. A simple way to determine $P_D$ is to calculate $V_{IN} \times I_{IN}$ when the output side is shorted. Input current gradually falls as temperature rises. You should use the value when thermal equilibrium is reached.

The range of usable currents can also be found from the graph below.



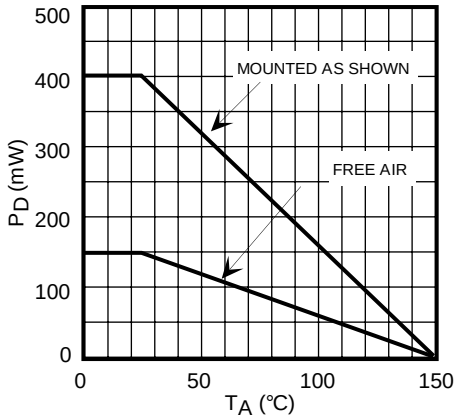

Procedure:

1) Find $P_D$

2) $P_{D1}$ is taken to be $P_D$ x (~ 0.8 - 0.9)
3) Plot $P_{D1}$ against 25 °C
4) Connect $P_{D1}$ to the point corresponding to the 150 °C with a straight line.
5) In design, take a vertical line from the maximum operating temperature (e.g., 75 °C) to the derating curve.
6) Read off the value of $P_D$ against the point at which the vertical line intersects the derating curve. This is taken as the maximum power dissipation, $D_{PD}$.

The maximum operating current is:

$$I_{OUT} = (D_{PD} / (V_{IN(MAX)} - V_{OUT})$$



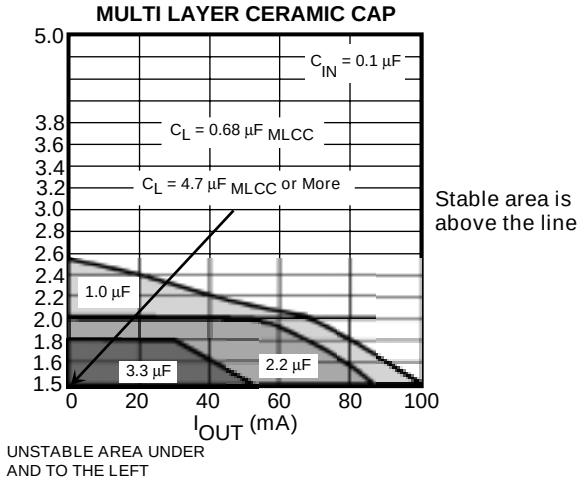

**SOT23-3 POWER DISSIPATION CURVE**

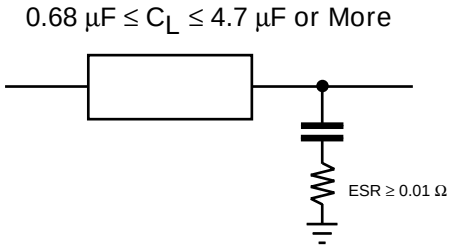
# APPLICATION INFORMATION

## INPUT-OUTPUT CAPACITORS

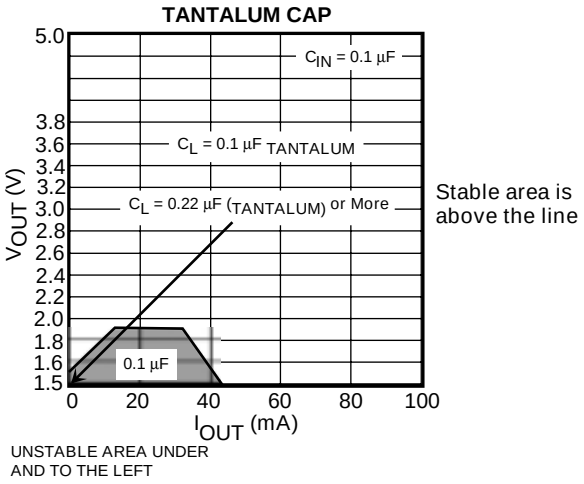
Linear regulators require input and output capacitors in order to maintain regulator loop stability. The output capacitor should be selected within the Equivalent Series Resistance (ESR) range as shown in the graphs below for stable operation. The output capacitor $C_L$ can be reduced as the output voltage is increased. However, the output noise will increase as $C_L$ is reduced, so the largest value of $C_L$ possilbe is recommended ($C_L$ = 4.7 $\mu$F or more).

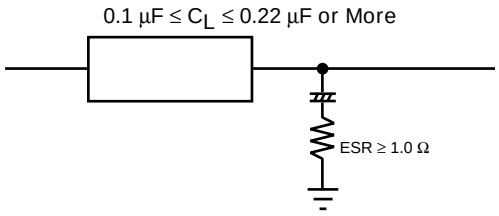
Note: It is very important to check the selected manufacturers electrical characteristics (capacitance and ESR) over temperature.




Range which can be used: $V_{OUT}$ ≥ 2.8V $I_{OUT}$ = 1 mA ~ Max, MULTI LAYER CERAMIC CAP. ≥ 0.68 $\mu$F
$V_{OUT}$ ≥ 2.0V $I_{OUT}$ = 1 mA ~ Max, MULTI LAYER CERAMIC CAP. ≥ 1.0 $\mu$F
$V_{OUT}$ ≥ 1.5V $I_{OUT}$ = 1 mA ~ Max, MULTI LAYER CERAMIC CAP. ≥ 4.7 $\mu$F




Range which can be used: $V_{OUT}$ ≥ 1.9V $I_{OUT}$ = 1 mA ~ Max, TANTALUM CAP. ≥ 0.1 $\mu$F (CERAMIC CAP 0.22 $\mu$F + 2.2Ω
Capacitor used for evaluation: $V_{OUT}$ ≥ 1.5V $I_{OUT}$ = 1 mA ~ Max, TANTALUM CAP. ≥ 0.22 $\mu$F (CERAMIC CAP 0.22 $\mu$F + 2.2Ω

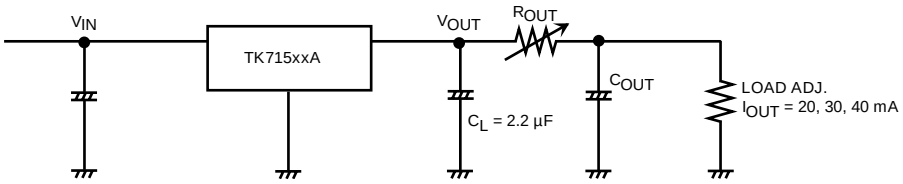
# APPLICATION INFORMATION (CONT)

### INPUT-OUTPUT CAPACITORS (CONT)

### OUTPUT NOISE IMPROVEMENT

An RC filter can be added to the output stage of the regulator to reduce output noise when the input voltage is high and the output current only makes small changes. Select a regulator with a slightly higher output voltage because the final output voltage will be reduced by the RC filter. If the output current does make a large change, the output voltage will change. The following table shows output noise, and output voltage for various values of $R_{OUT}$ and $C_{OUT}$ using a 3.5 V device.

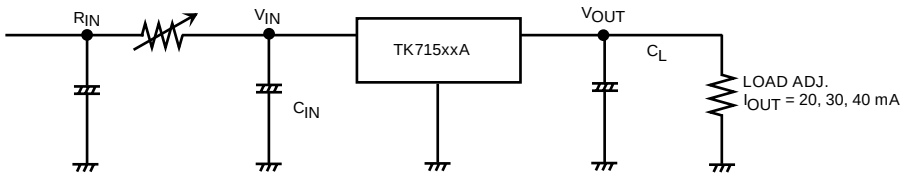
$V_{IN}$ = 4.0 V, $C_{IN}$ = 10 $\mu$F (aluminum electrolytic), $C_{OUT}$ = (see table), $R_{OUT}$ = (see table)




| MEASUREMENT CONDITION | | | OUTPUT SIDE CAPACITOR ($C_{OUT}$) | | | UNITS | |
|---|---|---|---|---|---|---|---|
| $I_{OUT}$ | $R_{OUT}$ | $V_{OUT}$ | 10 µF | 47 µF | 100 µF | µF | $C_{OUT}$ |
| $I_{OUT}$ = 20 - 40 mA | 0 | 3.500 V | 210 µV | 150 µV | 130 µV | RMS | NOISE |
| 20 mA | 10 Ω | 3.296 V | 76 µV | 50 µV | 40 µV | RMS | |
| 30 mA | 6.8 Ω | 3.287 V | 88 µV | 55 µV | 48 µV | RMS | |
| 40 mA | 5.1 Ω | 3.266 V | 100 µV | 60 µV | 48 µV | RMS | |

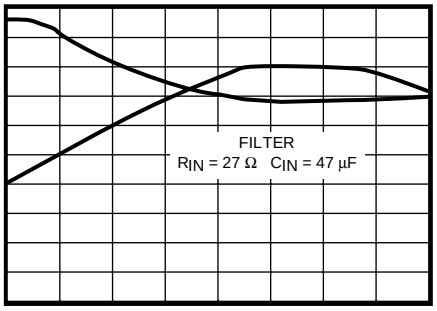
### RIPPLE REJECTION RATIO

An RC filter can be added to the input stage of the regulator to increase the ripple rejection when the input voltage is high. Even if the resulting difference between $V_{IN}$ at the regulator (after the RC filter) and $V_{OUT}$ is small the TK71533AS will output a stable voltage. The voltage dropped across the RC filter depends on the value of the input ripple noise. Select the value of $R_{IN}$ such that the lowest value of $V_{IN}$ plus the ripple noise after the RC filter (peak to peak) is the output voltage plus 0.2 V.



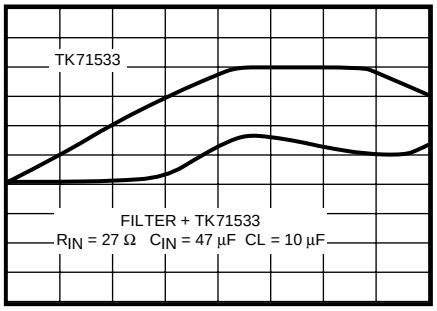
## APPLICATION INFORMATION (CONT)

R(*) MAG -17.98 dB 10 dB/ -50.00 dB
B(*) B -41.05 dB 10 dB/ -50.00 dB



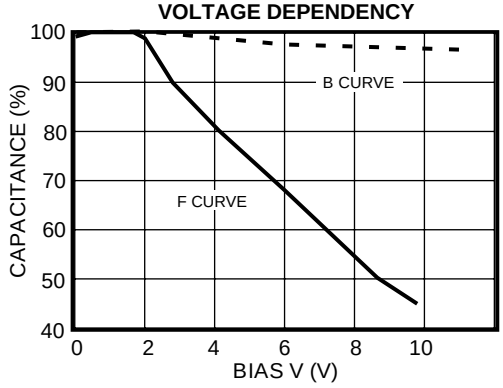

START: 100 Hz STOP: 1 MHz
OUT (B): -20.00 dBm ST: AUTO x1 1 MΩ
IRG: 26 dBm RBW: 30 kHZ VBW: 38 kHz

R(*) MAG -57.78 dB 10 dB/ -50.00 dB
B(*) B -41.05 dB 10 dB/ -50.00 dB

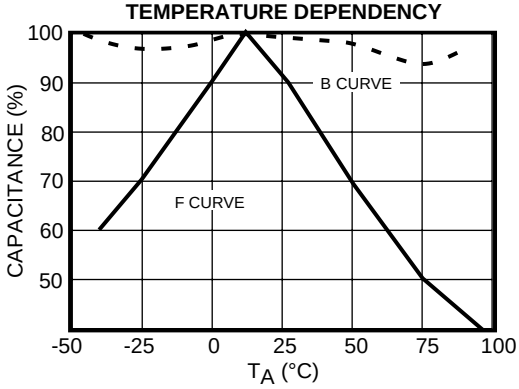



START: 100 Hz STOP: 1 MHz
OUT (B): -20.00 dBm ST: AUTO x1 1 MΩ
IRG: 26 dBm RBW: 30 kHZ VBW: 38 kHz

**VOLTAGE DEPENDENCY**




**TEMPERATURE DEPENDENCY**




In general, a ceramic capacitor has a voltage and temperature dependence. Parts should be selected with consideration of the voltage and temperature used. The “B” characteristic curves are recommended.

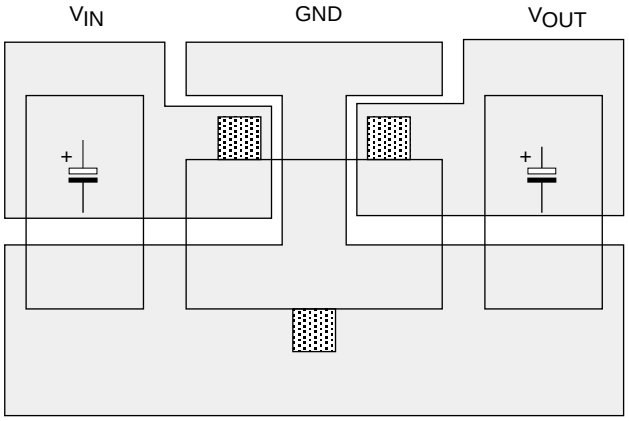
# APPLICATION INFORMATION (CONT.)

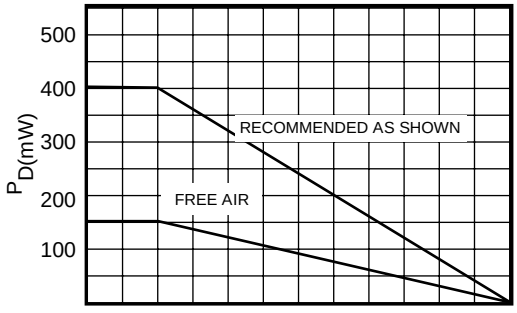
## BOARD LAYOUT

The copper pattern should be as large as possible. Power dissipation is 400 mW for the SOT23-3, derated at 3.2 mW/°C for operation above $T_A$ = 25°C ($\theta$ja = 312°C/W)

The internal reverse bias protection eliminates the requirement for a reverse voltage protection diode. This saves both cost and board space.

**SOT23-3 BOARD LAYOUT**

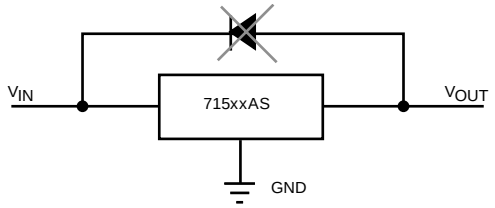


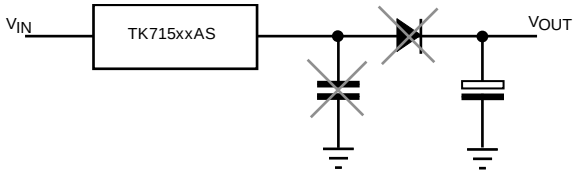
## APPLICATION INFORMATION (CONT.)

### REVERSE VOLTAGE PROTECTION

The internal reverse bias protection eliminates the requirement for a reverse voltage protection diode. This saves both cost and board space.

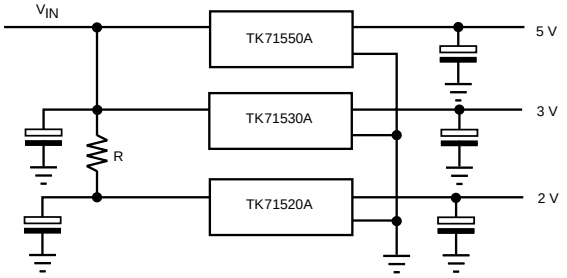



Another reverse voltage protection technique is illustrated below. The extra diode and extra capacitor are not necessary with the TK715xxA. The high output voltage accuracy is maintained because the diode forward voltage variations over temperature and load current have been eliminated.



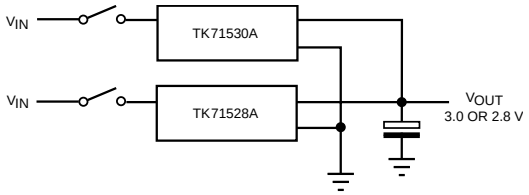

### PARALLEL OPERATION

The series resistor R is put in the input line of the low output voltage regulator in order to prevent overdissipation. The voltage dropped across the resistor reduces the large input-to-output voltage across the regulator, reducing the power dissipation in the device.



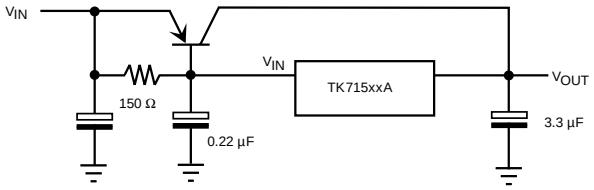

### SWITCHING OPERATION

Even though the input voltages or the output voltages are different, the outputs of the TK715xxA regulators can be connected together, and the output voltages switched. If two or more TK715xxA regulators are turned ON simultaneously, the highest output voltage will be present.



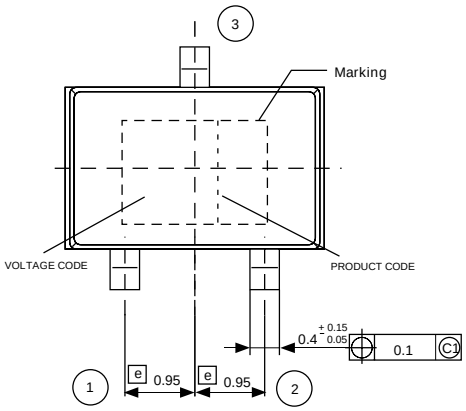

### CURRENT BOOST OPERATION

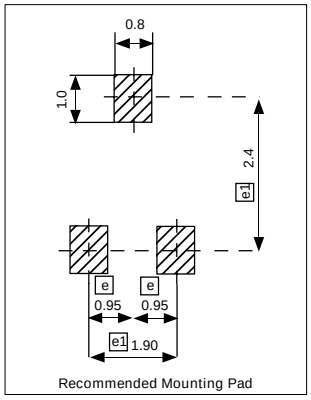
The output current can be increased by connecting an external PNP transistor as shown below. The output current capability depends upon the $H_{fe}$ of the external transistor. Note: The TK715xxA internal short circuit protection and thermal sensor do not protect the external transistor.



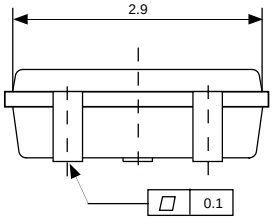
# PACKAGE OUTLINE

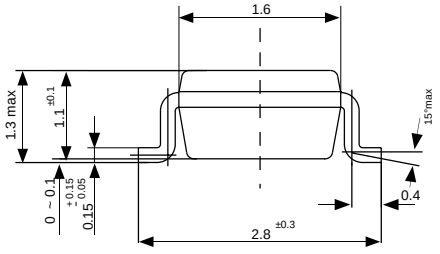
## SOT23-3




Recommended Mounting Pad




Dimensions are shown in millimeters
Tolerance: x.x = ± 0.2 mm (unless otherwise specified)

### Marking Information

| Product Code | T | Voltage Code |
|---|---|---|
| TK71515S | | 15 |
| TK71516S | | 16 |
| TK71517S | | 17 |
| TK71518S | | 18 |
| TK71519S | | 19 |
| TK71520S | | 20 |
| TK71521S | | 21 |
| TK71522S | | 22 |
| TK71523S | | 23 |
| TK71524S | | 24 |
| TK71525S | | 25 |
| TK71526S | | 26 |
| TK71527S | | 27 |
| TK71528S | | 28 |
| TK71529S | | 29 |
| TK71530S | | 30 |
| TK71531S | | 31 |
| TK71532S | | 32 |
| TK71533S | | 33 |
| TK71534S | | 34 |
| TK71535S | | 35 |
| TK71536S | | 36 |
| TK71537S | | 37 |
| TK71538S | | 38 |
| TK71539S | | 39 |
| TK71540S | | 40 |
| TK71541S | | 41 |
| TK71542S | | 42 |
| TK71543S | | 43 |
| TK71544S | | 44 |
| TK71545S | | 45 |
| TK71546S | | 46 |
| TK71547S | | 47 |
| TK71548S | | 48 |
| TK71549S | | 49 |
| TK71550S | | 50 |
| TK71560S | | 60 |
| TK71570S | | 70 |
| TK71580S | | 80 |
| TK71590S | | 90 |

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